

système 1^{er} ordre

* Equat différentielle

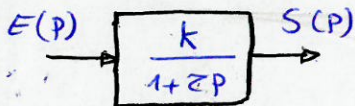
$$\tau \frac{ds(t)}{dt} + s(t) = k \cdot e(t)$$

* fnc t de transfert

$$H(p) = \frac{S(p)}{E(p)} = \frac{k}{1 + \tau p}$$

→ k: gain statique
→ τ: constante de temps

* schéma bloc



* Réponse indicielle

$$e(t) = E_0 \cdot u(t) \Rightarrow E(p) = \frac{E_0}{p}$$

$$\Rightarrow S(p) = E(p) \cdot \frac{k}{1 + \tau p} \Rightarrow S(p) = \frac{E_0}{p} \cdot \frac{k}{1 + \tau p}$$

⚠ d'après tableau de fnc s de transfert

$$s(t) = k E_0 (1 - e^{-t/\tau})$$

* Etude de la réponse indicielle

• le valeur finale

$$\lim_{t \rightarrow +\infty} s(t) = \lim_{p \rightarrow 0} p \cdot S(p) \Rightarrow \lim_{p \rightarrow 0} p \cdot \frac{E_0}{p} \cdot \frac{k}{1 + \tau p}$$

$$s_f = k \cdot E_0 \quad \text{avec} \quad k = \frac{S_{\infty}}{E_0}$$

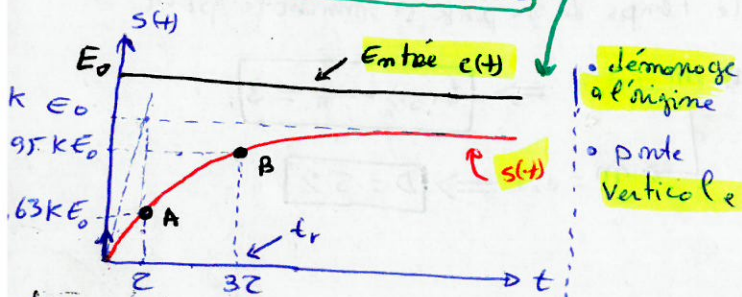
• le valeur initiale

$$\lim_{t \rightarrow 0} s(t) \Rightarrow \lim_{p \rightarrow +\infty} p \cdot S(p) \Rightarrow s_i = 0$$

le valeurs remarquable

• $t = \tau \Rightarrow s(t) = 63\% k E_0$

• $t = 3\tau \Rightarrow s(t) = 95\% k E_0$

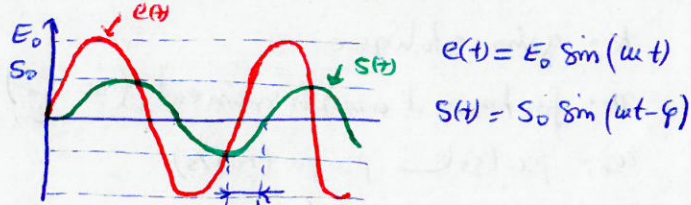
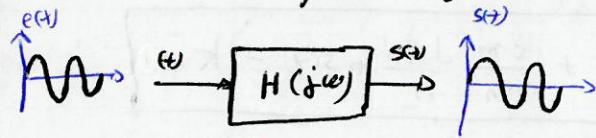


temps de réponse à 5% près

$$s(t_r) = 0.95 k E_0 = k E_0 (1 - e^{-t_r/\tau})$$

$$\Leftrightarrow e^{-t_r/\tau} = 0.05 \Rightarrow t_{r5\%} = 3\tau$$

* Réponse harmonique ou fréquentielle



* fnc de transfert complexe

$$p = j\omega \Rightarrow H(j\omega) = \frac{k}{1 + \tau j\omega}$$

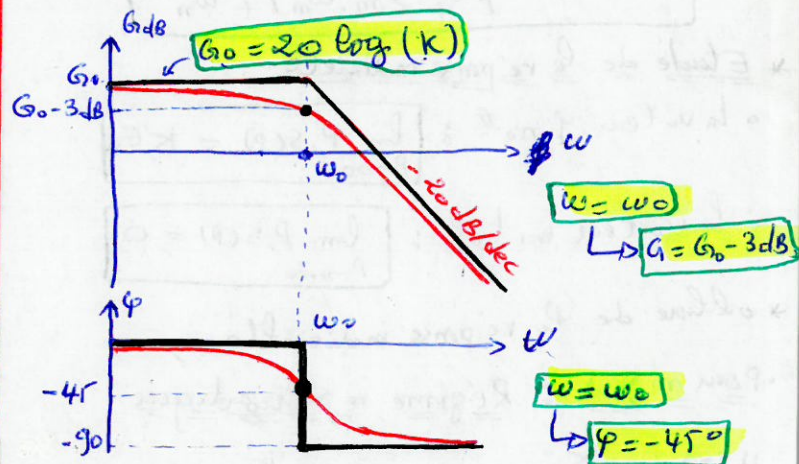
* module et phase de H(j*omega)

$$|H(j\omega)| = \frac{k}{\sqrt{1 + (\tau\omega)^2}}, \quad \varphi(j\omega) = -\arctg(\tau\omega)$$

* Diagramme de Bode

$$G_{dB} = 20 \log \left(\frac{k}{\sqrt{1 + (\tau\omega)^2}} \right)$$

$$G_{dB} = 20 \log(k) - 20 \log(\sqrt{1 + (\tau\omega)^2})$$



les infos qu'on peut tirer de D. Bode

• le gain statique: $k = 10^{\frac{G_0}{20}} \Leftrightarrow \omega = 0$

• la constante de temps: $\tau = \frac{1}{\omega_0}$

système 2^{ème} ordre

* équation de transfert $\frac{dx}{dt} \rightarrow P^2 X(P)$

$$\frac{1}{\omega_n^2} \frac{d^2 s(t)}{dt^2} + \frac{2m}{\omega_n} \frac{ds(t)}{dt} + s(t) = k e(t)$$

k : gain statique

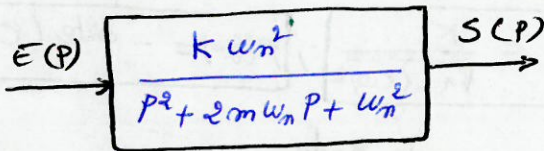
m : facteur d'amortissement (ou $\frac{\zeta}{2}$)

ω_n : pulsation propre (rad/s)

* fnct de transfert

$$H(P) = \frac{S(P)}{E(P)} = \frac{k \cdot \omega_n^2}{P^2 + 2m\omega_n P + \omega_n^2}$$

* schéma bloc



* Réponse indicielle

$$e(t) = E_0 \cdot u(t) \Rightarrow E(P) = \frac{E_0}{P}$$

$$\Rightarrow S(P) = \frac{E_0}{P} \frac{k \cdot \omega_n^2}{P^2 + 2m \cdot \omega_n \cdot P + \omega_n^2}$$

* Etude de la réponse indicielle

• la valeur finale : $\lim_{P \rightarrow 0} P \cdot S(P) = k E_0$

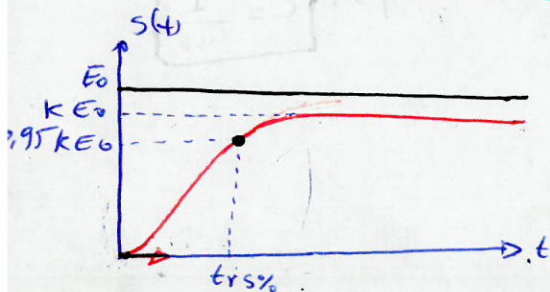
• la valeur initiale : $\lim_{P \rightarrow \infty} P \cdot S(P) = 0$

* obture de la réponse indicielle

* Pour $m > 1$: Régime aperiodique

$$H(P) = \frac{k}{\frac{1}{\omega_n^2} P^2 + \frac{2m}{\omega_n} P + 1} = \frac{k}{(1 + \tau_1 P)(1 + \tau_2 P)}$$

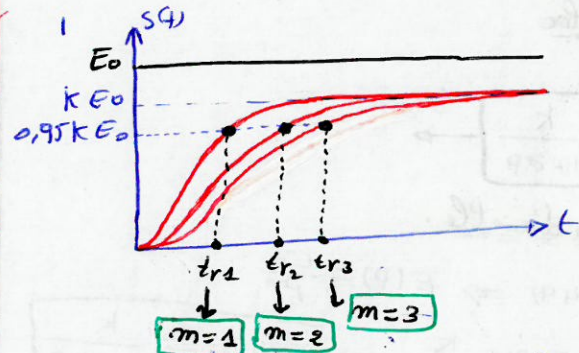
$$\text{si } E(P) = \frac{E_0}{P} \Rightarrow S(P) = \frac{E_0}{P} \frac{k}{(1 + \tau_1 P)(1 + \tau_2 P)}$$



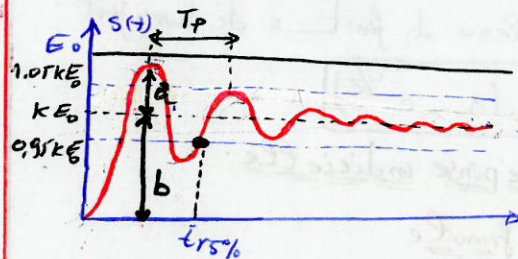
* Pour $m = 1$: Régime aperiodique critique

$$H(P) = \frac{k}{\frac{1}{\omega_n^2} P^2 + \frac{2m}{\omega_n} P + 1} = \frac{k}{(1 + \tau P)^2}$$

$$\text{si } E(P) = \frac{E_0}{P} \Rightarrow S(P) = \frac{E_0}{P} \frac{k}{(1 + \tau P)^2}$$

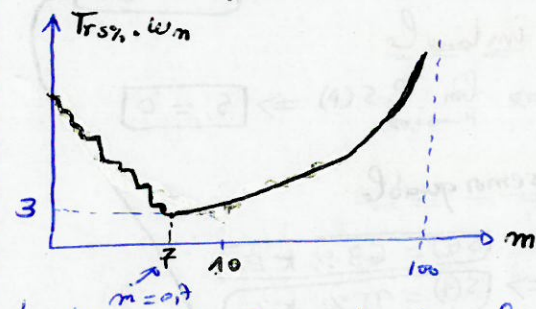


* Pour $m < 1$: régime oscillatoire amorti



• Dépossement : $D\% = 100 \times \frac{a}{b}$

• temps de réponse



Le temps de réponse est minimal pour

$m = 0.7 \Rightarrow trs\% \cdot \omega_n = 3$

$m = 0.7 \Rightarrow D = 5\%$

* Réponse harmonique 2^{ème} ordre réel

→ pour $m > 1$: deux solutions réelles

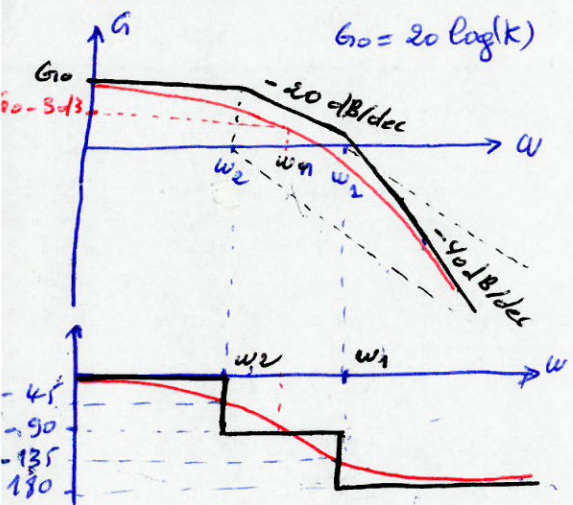
$$\begin{cases} \omega_1 = m\omega_n + \sqrt{m^2 - 1} \cdot \omega_n = \frac{1}{\tau_1} \\ \omega_2 = m\omega_n - \sqrt{m^2 - 1} \cdot \omega_n = \frac{1}{\tau_2} \end{cases}$$

$$H(j\omega) = \frac{k}{(1 + j\tau_1\omega)(1 + j\tau_2\omega)} \text{ avec } \omega_1 > \omega_2$$

$$|H(j\omega)| = \frac{k}{\sqrt{(1 + (\tau_1\omega)^2)(1 + (\tau_2\omega)^2)}}$$

$$\varphi(H(j\omega)) = -\arctg(\tau_1\omega) - \arctg(\tau_2\omega)$$

* Diagramme de Bode si $k > 0$



C/C:

a) $G_0 - 3 \text{ dB} \rightarrow \omega = \omega_n = \frac{1}{\sqrt{\tau_1\tau_2}}$

a) $\varphi = -45^\circ \rightarrow \omega = \omega_2 = \frac{1}{\tau_2}$

a) $\varphi = -90^\circ \rightarrow \omega = \omega_n$

a) $\varphi = -135^\circ \rightarrow \omega = \omega_1 = \frac{1}{\tau_1}$

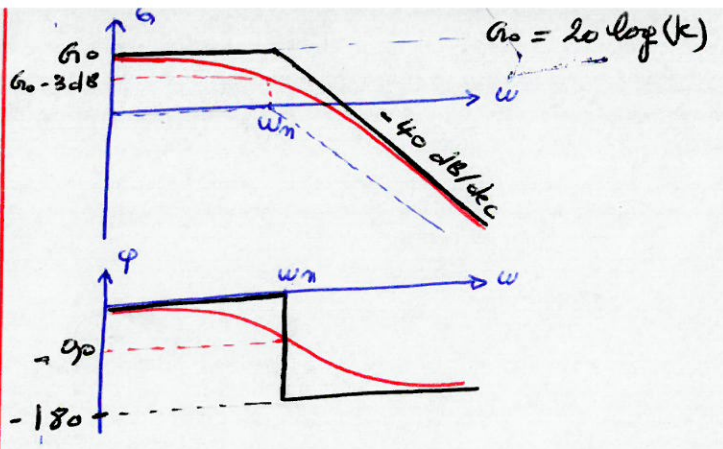
pour $m = 1 \Rightarrow$ deux solutions doubles

$$\omega_1 = \omega_2 = m\omega = \frac{1}{\tau} = \omega_0$$

$$H(j\omega) = \frac{k}{(1 + j\tau\omega)^2}$$

$$|H(j\omega)| = \frac{k}{1 + (\tau\omega)^2}$$

$$\varphi(H(j\omega)) = -2\arctg(\tau\omega)$$



C/C

- a) $G_0 - 3 \text{ dB} \rightarrow \omega = \omega_n = \frac{1}{\tau}$

- a) $\varphi = -90^\circ \rightarrow \omega = \omega_n = \frac{1}{\tau}$

* pour $m < 1$: deux solutions imaginaires

$$\begin{cases} \omega_1 = m\omega_n + j\sqrt{1 - m^2}\omega_n \\ \omega_2 = m\omega_n - j\sqrt{1 - m^2}\omega_n \end{cases}$$

$$H(j\omega) = \frac{k}{1 + 2m j \frac{\omega}{\omega_n} + (j \frac{\omega}{\omega_n})^2}$$

$$H(j\omega) = \frac{k}{1 - (\frac{\omega}{\omega_n})^2 + 2m j \frac{\omega}{\omega_n}}$$

$$|H(j\omega)| = \frac{k}{\sqrt{(1 - (\frac{\omega}{\omega_n})^2)^2 + (2m \frac{\omega}{\omega_n})^2}}$$

$$\varphi(j\omega) = -\arctg\left(\frac{2m \frac{\omega}{\omega_n}}{1 - (\frac{\omega}{\omega_n})^2}\right)$$

pour $k = 1$

